

## IV. CUADRIPOLI SI FILTRE ELECTRICE

### CAP. 13. CUADRIPOLI ELECTRICI

Breviar

a) *Forma fundamentala a ecuatiilor cuadripolilor si parametrii fundamentali: Prima forma fundamentala:*

$$\underline{U}_1 = \underline{A} \cdot \underline{U}_2 + \underline{B} \cdot \underline{I}_2$$

$$\underline{I}_1 = \underline{C} \cdot \underline{U}_2 + \underline{D} \cdot \underline{I}_2$$

A doua forma fundamentala:

$$\underline{U}_2 = \underline{D} \cdot \underline{U}_1 - \underline{B} \cdot \underline{I}_1$$

$$\underline{I}_2 = -\underline{C} \cdot \underline{U}_1 + \underline{A} \cdot \underline{I}_1$$

b) *Parametrii fundamentali au urmatoarele interpretari experimentale:*

$$\underline{A} = \left( \frac{\underline{U}_1}{\underline{U}_2} \right)_{\underline{I}_2=0} \quad \text{-- raportul de transformare al tensiunilor la functionarea in gol}$$

$$\underline{B} = \frac{1}{\left( \frac{\underline{I}_2}{\underline{U}_1} \right)_{\underline{U}_2=0}} \quad \text{-- valoarea reciproca a admitantei de transfer la functionarea in scurtcircuit;}$$

$$\underline{C} = \frac{1}{\left( \frac{\underline{U}_2}{\underline{I}_1} \right)_{\underline{I}_2=0}} \quad \text{-- valoarea reciproca a impedantei de transfer la functionarea in gol;}$$

$$\underline{D} = \left( \frac{\underline{I}_1}{\underline{I}_2} \right)_{\underline{U}_2=0} \quad \text{-- raportul de transformare al curentilor la functionarea in scurtcircuit.}$$

c) *Marimi caracteristice ale cuadripolilor:*

– impedanta primara de mers in gol:

$$\underline{Z}_{10} = \left( \frac{\underline{U}_1}{\underline{I}_1} \right)_{\underline{I}_2=0} = \frac{\underline{A} \cdot \underline{U}_{20}}{\underline{C} \cdot \underline{U}_{20}} = \frac{\underline{A}}{\underline{C}}$$

– impedanta primara de scurtcircuit:

$$\underline{Z}_{isc} = \left( \frac{\underline{U}_1}{\underline{I}_1} \right)_{\underline{U}_2=0} = \frac{\underline{B}}{\underline{D}}$$

– impedanta secundara de mers in gol:

$$\underline{Z}_{20} = \left( \frac{\underline{U}'_2}{\underline{I}'_2} \right)_{\underline{I}'_1=0} = \frac{\underline{D} \cdot \underline{U}'_{10}}{\underline{C} \cdot \underline{U}'_{10}} = \frac{\underline{D}}{\underline{C}}$$

– impedanta secundara de scurtcircuit:

$$\underline{Z}_{2sc} = \left( \frac{\underline{U}'_2}{\underline{I}'_2} \right)_{\underline{U}'_1=0} = \frac{\underline{B} \cdot \underline{I}'_1}{\underline{A} \cdot \underline{I}'_1} = \frac{\underline{B}}{\underline{A}}$$

d) Parametrii impedanta si cuadripolului:

Daca parametrul fundamental  $\underline{C}$  este diferit de zero ecuatiile fundamentale ale cuadripolului se pot explicita in raport cu tensiunile  $\underline{U}_1$ ,  $\underline{U}_2$  si se obtine urmatoarea forma a ecuatiilor cuadripolilor liniari si pasivi:

$$\underline{U}_1 = \underline{Z}_{11} \cdot \underline{I}_1 + \underline{Z}_{12} \cdot \underline{I}_2 \text{ si}$$

$$\underline{U}_2 = \underline{Z}_{21} \cdot \underline{I}_1 + \underline{Z}_{22} \cdot \underline{I}_2, \text{ unde:}$$

$$\underline{Z}_{11} = \left( \frac{\underline{U}_1}{\underline{I}_1} \right)_{\underline{I}_2=0} = \frac{\underline{A}}{\underline{C}}$$

$$\underline{Z}_{12} = \left( \frac{\underline{U}_1}{\underline{I}_2} \right)_{\underline{I}_1=0} = \frac{\underline{A} \cdot \underline{D} - \underline{B} \cdot \underline{C}}{\underline{C}}$$

$$\underline{Z}_{21} = \left( \frac{\underline{U}_2}{\underline{I}_1} \right)_{\underline{I}_2=0} = \frac{1}{\underline{C}};$$

$$\underline{Z}_{22} = \left( \frac{\underline{U}_2}{\underline{I}_2} \right)_{\underline{I}_1=0} = -\frac{\underline{D}}{\underline{C}}$$

e) Parametrii admitanta ai cuadripolului.

Daca parametrul fundamental  $\underline{B}$  este diferit de zero, in ecuatiile fundamentale se pot explicita curentii  $\underline{I}_1, \underline{I}_2$ , obtinându-se sistemul de ecuatii:

$$\underline{I}_1 = \underline{Y}_{11} \underline{U}_1 + \underline{Y}_{12} \underline{U}_2$$

$$\underline{I}_2 = \underline{Y}_{21} \underline{U}_1 + \underline{Y}_{22} \underline{U}_2$$

$$\underline{Y}_{11} = \left( \frac{\underline{I}_1}{\underline{U}_1} \right)_{\underline{U}_2=0} = \frac{\underline{D}}{\underline{B}}$$

$$\underline{Y}_{12} = \left( \frac{\underline{I}_1}{\underline{U}_2} \right)_{\underline{U}_1=0} = \frac{\underline{A} \cdot \underline{D} - \underline{B} \cdot \underline{C}}{\underline{B}}$$

$$\underline{Y}_{21} = \left( \frac{\underline{I}_2}{\underline{U}_1} \right)_{\underline{U}_2=0} = \frac{1}{\underline{B}}$$

$$\underline{Y}_{22} = \left( \frac{\underline{I}_2}{\underline{U}_2} \right)_{\underline{U}_1=0} = -\frac{\underline{A}}{\underline{B}}$$

13.1. (R) Sa se determine constantele  $\underline{A}, \underline{B}, \underline{C}, \underline{D}$  si impedanta caracteristica primara pentru cuadripolul din fig.13.1.a. Date numerice:  $R=10\Omega$ ;  $R_0=5\Omega$ ;  $X=10\Omega$ ;  $X_0=5\Omega$ .

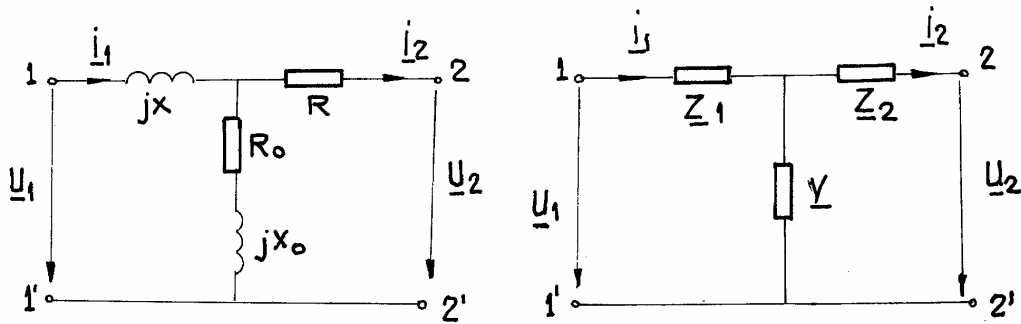


Fig. 13.1

*Rezolvare:* Cuadripolul dat fiind un cuadripol in T, se va considera schema generala (fig.13.1.b) a unui cuadripol T si se vor determina pentru aceasta schema constantele A, B, C, D.

Aplicând metoda curentilor ciclici se obtin ecuatiile:

$$\begin{aligned} \left( \underline{Z}_1 + \frac{1}{\underline{Y}} \right) \cdot \underline{I}_1 - \frac{1}{\underline{Y}} \cdot \underline{I}_2 &= \underline{U}_1 \\ -\frac{1}{\underline{Y}} \cdot \underline{I}_1 + \left( \underline{Z}_2 + \frac{1}{\underline{Y}} \right) \underline{I}_2 &= -\underline{U}_2 \end{aligned}$$

deoarece  $\underline{I}_1 = \underline{I}_1; \underline{I}_2 = \underline{I}_2$

$$\text{Din ultima ecuatie: } \underline{I}_1 = \underline{Y} \cdot \underline{U}_2 + (1 + \underline{Z}_2 \cdot \underline{Y}) \cdot \underline{I}_2$$

si, introducând in prima, se obtine:

$$\underline{U}_1 = (1 + \underline{Z}_1 \cdot \underline{Y}) \cdot \underline{U}_2 + (\underline{Z}_1 + \underline{Z}_2 + \underline{Z}_1 \cdot \underline{Z}_2 \cdot \underline{Y}) \cdot \underline{I}_2$$

Prin urmare:

$$\underline{A} = 1 + \underline{Z}_1 \cdot \underline{Y}$$

$$\underline{B} = \underline{Z}_1 + \underline{Z}_2 + \underline{Z}_1 \underline{Z}_2 \cdot \underline{Y}$$

$$\underline{C} = \underline{Y}$$

$$\underline{D} = 1 + \underline{Z}_2 \cdot \underline{Y}$$

numeric, se gaseste:

$$\underline{Z}_1 = j \cdot X = j \cdot 10 \Omega$$

$$\underline{Z}_2 = R = 10 \Omega$$

$$\underline{Y} = \frac{1}{R_0 + j \cdot X_0} = \frac{1}{5 + 5j}$$

Deci:

$$\underline{A} = 1 + \underline{Z}_1 \cdot \underline{Y} = 1 + \frac{j \cdot 10}{5 + j \cdot 5} = 2 + j$$

$$\underline{B} = \underline{Z}_1 + \underline{Z}_2 + \underline{Z}_1 \cdot \underline{Z}_2 \cdot \underline{Y} = j \cdot 10 + 10 + j \cdot 10 \cdot 10 \cdot \frac{1}{5 + j \cdot 5} = (20 + j \cdot 20) \Omega$$

$$\underline{C} = \underline{Y} = \frac{1}{5 + j \cdot 5} = (0,1 - j \cdot 0,1) S$$

$$\underline{D} = 1 + \underline{Z}_2 \cdot \underline{Y} = 1 + 10 \cdot \frac{1}{5 + j \cdot 5} = 1 + 1 - j = 2 - j$$

Pentru calculul impedantei caracteristice primare se foloseste relatia:

$$\underline{Z}_{1c} = \frac{\underline{A} \cdot \underline{Z}_{1c} + \underline{B}}{\underline{C} \cdot \underline{Z}_{1c} + \underline{D}}$$

din care rezulta ecuatia:

$$\underline{C} \cdot \underline{Z}_{1c}^2 + (\underline{D} - \underline{A}) \cdot \underline{Z}_{1c} - \underline{B} = 0$$

pe care rezolvând-o se obtine:

$$\underline{Z}_{1c} = \frac{\underline{A} - \underline{D}}{2\underline{C}} \pm \sqrt{\frac{(\underline{A} - \underline{D})^2}{4\underline{C}^2} + \frac{\underline{B}}{\underline{C}}}$$

si numeric:

$$\begin{aligned} \underline{Z}_{1c} &= \frac{2 + j - 2 + j}{2(0,1 - j \cdot 0,1)} \pm \sqrt{\frac{(2j)^2}{4(0,1 - j \cdot 0,1)^2} + \frac{20 + j \cdot 20}{0,1 - j \cdot 0,1}} = \\ &= \begin{cases} 5(\sqrt{3} - 1) + j \cdot 5(1 + \sqrt{3}) \quad \Omega \\ -5(\sqrt{3} + 1) - j \cdot 5(\sqrt{3} - 1) \quad \Omega \end{cases} \end{aligned}$$

Se retine numai prima valoare deoarece a doua are partea reala negativa si deci nu poate fi realizabila practic, deci:

$$\underline{Z}_{1c} = 5(\sqrt{3} - 1) + j \cdot 5(\sqrt{3} + 1) = (3,65 + j \cdot 13,65) \Omega$$

adica este formata dintr-o rezistenta de  $3,65 \Omega$  in serie cu o bobina ideala de inductivitate:

$$L = \frac{13,65}{2 \cdot \pi \cdot 50} = \frac{13,65}{314} = 4,34 \cdot 10^{-2} \text{ H} = 43,4 \text{ mH in cazul frecventei de alimentare}$$

f=50 Hz.

13.2 (R) Sa se determine valorile elementelor schemelor echivalente in T,  $\pi$  si  $\Gamma$  ale unui cuadripol oarecare ai carui parametri  $\underline{A}$ ,  $\underline{C}$ ,  $\underline{D}$  au valorile:  $\underline{A} = 0,3 + j \cdot 0,6$ ;  $\underline{C} = (0,1 + j \cdot 0,2)$ ;  $\underline{D} = 0,5 + j \cdot 0,5$ .

*Rezolvare:* Aplicând teorema a II-a lui Kirchhoff celor doua ochiuri independente ale schemei echivalente in T din fig.13.2.a, se obtin ecuatiile:

$$\underline{U}_1 = \underline{Z}_1 \cdot \underline{I}_1 + \frac{1}{\underline{Y}} (\underline{I}_1 - \underline{I}_2)$$

$$-\underline{U}_2 = \underline{Z}_2 \cdot \underline{I}_2 - \frac{1}{\underline{Y}} (\underline{I}_1 - \underline{I}_2)$$

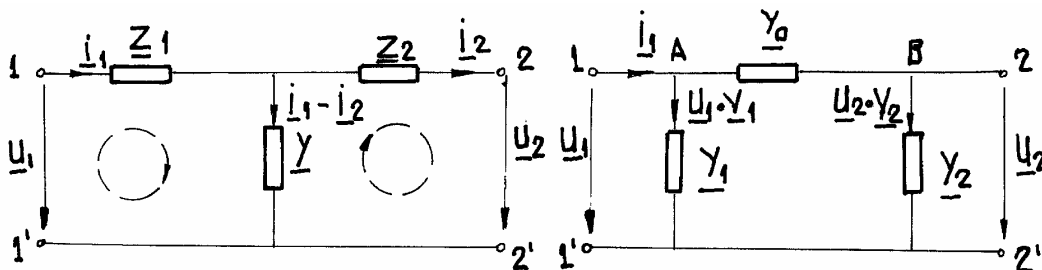


Fig. 13.2

Rezolvând acest sistem de ecuatii in raport cu marimile de la intrare ( $\underline{I}_1, \underline{U}_1$ ) se obtin ecuatiile:

$$\underline{I}_1 = \underline{U}_2 \cdot \underline{Y} + (\underline{Z}_2 \cdot \underline{Y} + 1) \cdot \underline{I}_2$$

$$\underline{U}_1 = (\underline{Z}_1 \cdot \underline{Y} + 1) \underline{U}_2 + (\underline{Z}_1 + \underline{Z}_2 + \underline{Z}_1 \cdot \underline{Z}_2 \cdot \underline{Y}) \underline{I}_2$$

de unde, prin comparatie cu ecuatiile:

$$\underline{U}_1 = \underline{A} \cdot \underline{U}_2 + \underline{B} \cdot \underline{I}_2$$

$$\underline{I}_1 = \underline{C} \cdot \underline{U}_2 + \underline{D} \cdot \underline{I}_2$$

se obtin coeficientii fundamentali:

$$\underline{A} = 1 + \underline{Z}_1 \cdot \underline{Y};$$

$$\underline{B} = \underline{Z}_1 \cdot \underline{Z}_2 \cdot \underline{Y} + \underline{Z}_1 + \underline{Z}_2;$$

$$\underline{C} = \underline{Y};$$

$$\underline{D} = 1 + \underline{Z}_2 \cdot \underline{Y}$$

Rezulta de aici:

$$\underline{Y} = \underline{C}; \quad \underline{Z}_1 = (\underline{A} - 1) / \underline{C}; \quad \underline{Z}_2 = (\underline{D} - 1) / \underline{C}$$

Numeric:

$$\underline{Z} = 1 / \underline{Y} = (2 - j4) \Omega; \quad \underline{Z}_1 = (1 + j4) \Omega; \quad \underline{Z}_2 = (1 + j3) \Omega$$

In cazul cuadripolului in  $\pi$ , se aplica prima teorema a lui Kirchhoff la nodul B si a doua teorema a lui Kirchhoff pe ochiul care se inchide in lungul celor doua tensiuni la borne (fig.13.2.b) si se obtin ecuatiile:

$$\underline{I}_1 - \underline{U}_1 \cdot \underline{Y}_1 = \underline{I}_2 + \underline{U}_2 \cdot \underline{Y}_2$$

$$\underline{U}_1 = \frac{1}{\underline{Y}_0} (\underline{I}_2 + \underline{U}_2 \cdot \underline{Y}_2) + \underline{U}_2$$

sau:

$$\underline{U}_1 = (\underline{Y}_2 / \underline{Y}_0 + 1) \cdot \underline{U}_2 + 1 / \underline{Y}_0 \cdot \underline{I}_2$$

$$\underline{I}_1 = \left( \underline{Y}_1 + \underline{Y}_2 + \frac{\underline{Y}_1 \underline{Y}_2}{\underline{Y}_0} \right) \cdot \underline{U}_2 + \left( \frac{\underline{Y}_1}{\underline{Y}_0} + 1 \right) \underline{I}_2$$

de unde rezulta:

$$\underline{A} = 1 + \frac{\underline{Y}_2}{\underline{Y}_0}; \quad \underline{B} = 1 / \underline{Y}_0$$

$$\underline{C} = \underline{Y}_1 + \underline{Y}_2 + \frac{\underline{Y}_1 \cdot \underline{Y}_2}{\underline{Y}_0}; \quad \underline{D} = 1 + \underline{Y}_1 / \underline{Y}_0$$

sau:

$$\underline{Y}_0 = 1 / \underline{B}; \quad \underline{Y}_1 = (\underline{D} - 1) / \underline{B}; \quad \underline{Y}_2 = (\underline{A} - 1) / \underline{B}$$

Numeric:

$$\underline{B} = \frac{\underline{A} \cdot \underline{D} - 1}{\underline{C}} = (-0,5 + j \cdot 5,5) \Omega$$

$$\underline{Y}_0 = (-0,0164 + j \cdot 0,180) \text{ S}$$

$$\underline{Y}_1 = (0,0985 + j \cdot 0,082) \text{ S}$$

$$\underline{Y}_2 = (0,12 + j \cdot 0,116) \text{ S}$$

**Observatii:**

a) Cuadripolul echivalent in T poate fi realizat (impedantele respective au partile reale pozitive) sub forma din fig.13.2.1;

b) Cuadripolul echivalent in  $\pi$  nu poate fi realizat (deoarece  $Y_0$  are partea reala negativa).

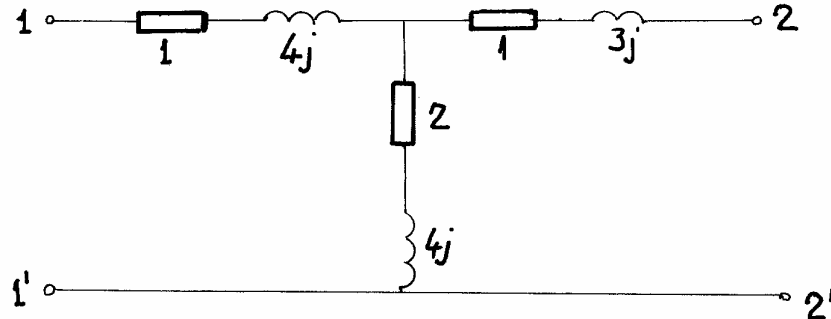


Fig. 13.2.1

Pentru determinarea elementelor cuadripolului echivalent in  $\Gamma$  (fig.13.2.2) se pot scrie ecuatiile:

$$\underline{I}_1 = \underline{I}_2 + \underline{Y} \cdot \underline{U}_2$$

$$\underline{U}_1 = \underline{U}_2 + \underline{Z} \cdot \underline{I}_1$$

sau:

$$\underline{U}_1 = \underline{U}_2 + \underline{Z}(\underline{I}_2 + \underline{Y} \cdot \underline{U}_2) = (1 + \underline{Z} \cdot \underline{Y}) \cdot \underline{U}_2 + \underline{Z} \cdot \underline{I}_2$$

$$\underline{I}_1 = \underline{Y} \cdot \underline{U}_2 + \underline{I}_2$$

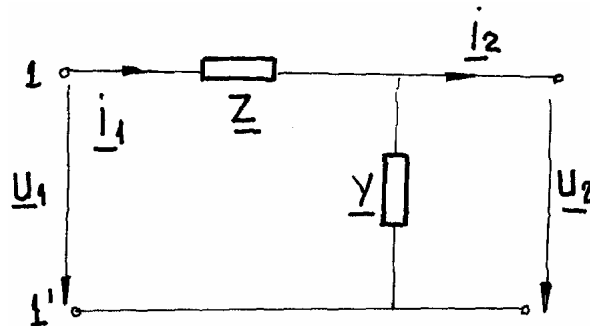


Fig. 13.2.2

Identificând cu ecuatiile:

$$\underline{U}_1 = \underline{A} \cdot \underline{U}_2 + \underline{B} \cdot \underline{I}_2 \text{ se obtine: } \underline{A} = 1 + \underline{Z} \cdot \underline{Y}; \underline{B} = \underline{Z}$$

$$\underline{I}_1 = \underline{C} \cdot \underline{U}_2 + \underline{D} \cdot \underline{I}_2 \quad \underline{C} = \underline{Y}; \underline{D} = 1$$

$$\text{Deci: } \underline{Z} = \underline{B} = (0,5 + j \cdot 5,5) \Omega$$

$$\underline{Y} = \underline{C} = (0,1 + j \cdot 0,2) \text{ S}$$

$$\underline{Z}_0 = 1 / \underline{Y} = 1 / (0,1 + j \cdot 0,2) = 10 / (1 + 2j) = (2 - 4j) \Omega$$

realizarea acestui cuadripol este aratata in fig.13.2.3:

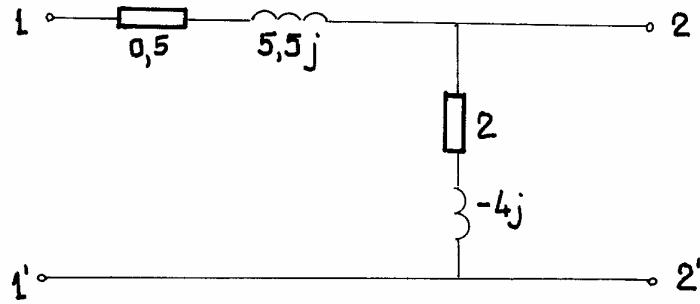


Fig. 13.2.3

13.3. (R) Pentru cuadripolul din fig.13.3, sa se determine coeficientii  $\underline{Z}_{11}, \underline{Z}_{12}, \underline{Z}_{22}$ , coeficientii  $\underline{A}, \underline{B}, \underline{C}, \underline{D}$  precum si elementele cuadripolului in T echivalent. Date numerice:  $X_1 = \omega \cdot L_1 = 3\Omega$ ,  $X_2 = \omega \cdot L_2 = 2\Omega$ ;  $R = 1\Omega$ ,  $X_M = \omega M = 1\Omega$ .

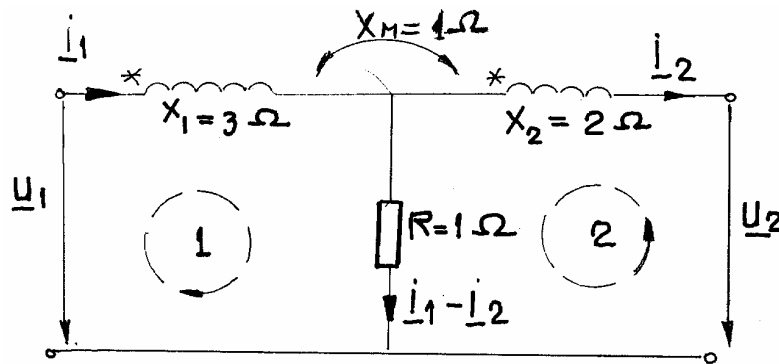


Fig.13.3

*Rezolvare:* Aplicând teorema a doua a lui Kirchoff pe ochiurile 1 si 2, cu sensurile de parcurgere pozitive a ochiurilor alese ca in figura, se obtine sistemul de ecuatii:

$$\begin{cases} \underline{U}_1 = j \cdot X_1 \cdot \underline{I}_1 + R(\underline{I}_1 - \underline{I}_2) + j \cdot X_M \cdot \underline{I}_2 \\ \underline{U}_2 = -j \cdot X_2 \cdot \underline{I}_2 + R(\underline{I}_1 - \underline{I}_2) - j \cdot X_M \cdot \underline{I}_1 \end{cases}$$

sau:

$$\begin{cases} \underline{U}_1 = (R + j \cdot X_1) \cdot \underline{I}_1 + (-R + j \cdot X_M) \cdot \underline{I}_2 \\ \underline{U}_2 = (R - j \cdot X_M) \cdot \underline{I}_1 + (-R - j \cdot X_2) \cdot \underline{I}_2 \end{cases}$$

Rezulta:  $\underline{Z}_{11} = R + j \cdot X_1$ ;  $\underline{Z}_{12} = -\underline{Z}_{21} = -R + j \cdot X_M$ ;  
 $\underline{Z}_{22} = -R - j \cdot X_2$

sau numeric:  $\underline{Z}_{11} = (1 + j \cdot 3)\Omega$ ;  $\underline{Z}_{12} = (-1 + j)\Omega$ ;  $\underline{Z}_{22} = (-1 - j2)\Omega$ .

Rezolvând cea de-a doua ecuație a sistemului de mai sus în raport cu  $\underline{I}_1$  se obține:

$$\underline{I}_1 = \frac{1}{1-j} \underline{U}_2 + \frac{1+2j}{1-j} \underline{I}_2$$

adică:

$$\underline{C} = 1/(1-j) = (0,5 + j \cdot 0,5) \text{ S}$$

$$\underline{D} = (1+2j)/(1-j) = -0,5 + j \cdot 1,5$$

Pentru determinarea coeficienților  $\underline{A}$  și  $\underline{B}$  se poate exprima  $\underline{U}_1$  în funcție de  $\underline{U}_2$  și de  $\underline{I}_2$  sau se folosesc formulele:

$$\underline{A} = \left( \frac{\underline{U}_1}{\underline{U}_2} \right)_{\underline{I}_2=0} = \frac{\underline{Z}_{11}}{\underline{Z}_{21}}; \quad \underline{C} = 1/\underline{Z}_{21};$$

$$\underline{B} = (\underline{A} \cdot \underline{D} - 1)/\underline{C} = \underline{Z}_{12} - (\underline{Z}_{11} \cdot \underline{Z}_{22})/\underline{Z}_{21}$$

$$\underline{D} = -\underline{Z}_{22}/\underline{Z}_{21}$$

Înlocuind numeric, se obține:

$$\underline{A} = \frac{1+j3}{1-j} = -1+j2; \quad \underline{B} = \frac{(-1+2j)(-0,5+j \cdot 1,5)}{0,5+j \cdot 0,5} = (-6+j)\Omega$$

Elementele cuadripolului în T echivalent sunt cf. fig. 13.3.1.

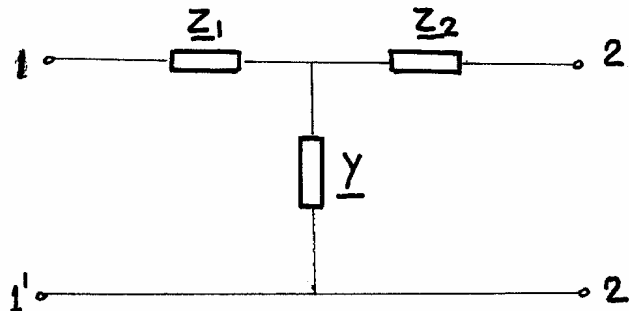


Fig. 13.3.1

$$\underline{Y} = \underline{C} = (0,5 + j \cdot 0,5) \text{ S}$$

$$\underline{Z} = 1/\underline{Y} = 1/(0,5 + j \cdot 0,5) = (1-j) \Omega$$

$$\underline{Z}_1 = (\underline{A} - 1)/\underline{C} = (-1 + j \cdot 2 - 1)/(0,5 + j \cdot 0,5) = 4j \Omega$$

$$\underline{Z}_2 = (\underline{D} - 1)/\underline{C} + (-0,5 + j \cdot 1,5 - 1)/(1+j)/2 = 3j \Omega$$

Realizarea cuadripolului este:



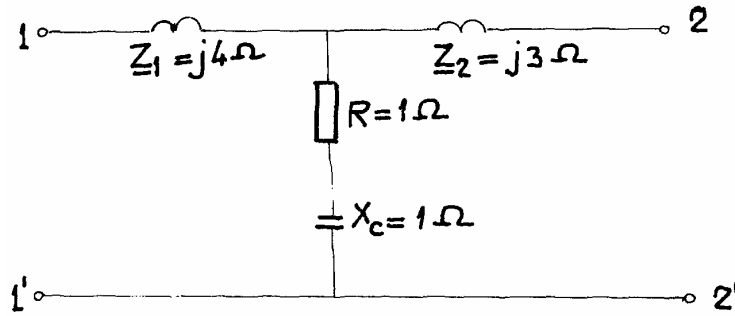


Fig. 13.3.2

13.4. (R) Sa se determine parametrii  $\underline{A}, \underline{B}, \underline{C}, \underline{D}$  si  $\underline{Y}_{11}, \underline{Y}_{12}, \underline{Y}_{22}$  ai cuadripolului din fig.13.4 si elementele cuadripolului in  $\pi$  echivalent. Date numerice:  $R=1\Omega; X_1=2\Omega; X_2=2\Omega, X_M=1\Omega$ .

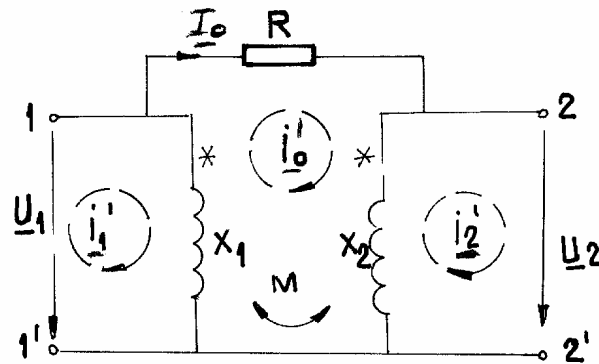


Fig. 13.4

*Rezolvare:* Aplicând metoda curenților ciclici (fig.13.4), se obține sistemul de ecuații:

$$\begin{aligned} \underline{U}_1 &= j \cdot X_1 \cdot \underline{I}'_1 - j \cdot X_M \cdot \underline{I}'_2 + j(X_M - X_1) \cdot \underline{I}'_0 \\ -\underline{U}_2 &= -j \cdot X_M \cdot \underline{I}'_1 + j \cdot X_2 \cdot \underline{I}'_2 + j(X_M - X_2) \cdot \underline{I}'_0 \\ 0 &= j(X_M - X_1) \cdot \underline{I}'_1 + j(X_M - X_2) \cdot \underline{I}'_2 + [R + j(X_1 + X_2 - 2X_M)] \cdot \underline{I}'_0 \end{aligned}$$

Inlocuind datele numerice si tinând cont ca  $\underline{I}_1 = \underline{I}'_1; \underline{I}_2 = \underline{I}'_2; \underline{I}_0 = \underline{I}'_0$  sistemul devine:

$$\begin{aligned} \underline{U}_1 &= 2j \cdot \underline{I}_1 - j \cdot \underline{I}_2 - j \cdot \underline{I}_0 \\ \underline{U}_2 &= j \cdot \underline{I}_1 - 2j \cdot \underline{I}_2 + j \cdot \underline{I}_0 \\ 0 &= -j \cdot \underline{I}_2 - j \cdot \underline{I}_1 + (1 + 2j) \cdot \underline{I}_0 \end{aligned}$$

de unde, eliminând pe  $\underline{I}_0$  si explicitând curenții  $\underline{I}_1$  si  $\underline{I}_2$  se obțin ecuațiile:

$$\begin{aligned} \underline{I}_1 &= \frac{3-2j}{3} \underline{U}_1 + \frac{-3+j}{3} \underline{U}_2 \\ \underline{I}_2 &= \frac{3-j}{3} \underline{U}_1 + \frac{-3+2j}{3} \underline{U}_2 \end{aligned}$$

Deci:

$$\underline{Y}_{11} = \frac{3-2j}{3} \text{S}; \underline{Y}_{12} = -\underline{Y}_{21} = \frac{-3+j}{3}; \underline{Y}_{22} = \frac{-3+2j}{3} \text{S}$$

Pentru determinarea coeficientilor  $\underline{A}$ ,  $\underline{B}$ ,  $\underline{C}$ ,  $\underline{D}$  se explicita sistemul de mai sus in raport cu  $\underline{U}_1$  si  $\underline{I}_1$  si se obtine:

$$\underline{U}_1 = \frac{11-3j}{10} \underline{U}_2 + \frac{9+3j}{10} \underline{I}_2$$

$$\underline{I}_1 = -\frac{1+7j}{10} \underline{U}_2 + \frac{11-3j}{10} \underline{I}_2$$

Prin urmare:

$$\underline{A} = \frac{11-3j}{10}; \underline{B} = \frac{9+3j}{10} \Omega; \underline{C} = -\frac{1+7j}{10} S; \underline{D} = \frac{11-3j}{10}$$

Elementele cuadripolului echivalent in  $\pi$  se determina scriind ecuatiile care rezulta din aplicarea primei teoreme a lui Kirchoff in nodurile M si N (fig. 13.4.1.a):

$$\underline{I}_1 = (\underline{U}_1 - \underline{U}_2) \cdot \underline{Y}'_0 + \underline{U}_1 \cdot \underline{Y}'_1 = (\underline{Y}'_1 + \underline{Y}'_0) \cdot \underline{U}_1 - \underline{Y}'_0 \cdot \underline{U}_2$$

$$\underline{I}_2 = (\underline{U}_1 - \underline{U}_2) \cdot \underline{Y}'_0 - \underline{U}_2 \cdot \underline{Y}'_2 = \underline{Y}'_0 \cdot \underline{U}_1 - (\underline{Y}'_0 + \underline{Y}'_2) \cdot \underline{U}_2$$

sau:

$$\underline{Y}'_{11} = \underline{Y}'_1 + \underline{Y}'_0; \underline{Y}'_{22} = -(\underline{Y}'_2 + \underline{Y}'_0); \underline{Y}'_{12} = -\underline{Y}'_{21} = -\underline{Y}'_0$$

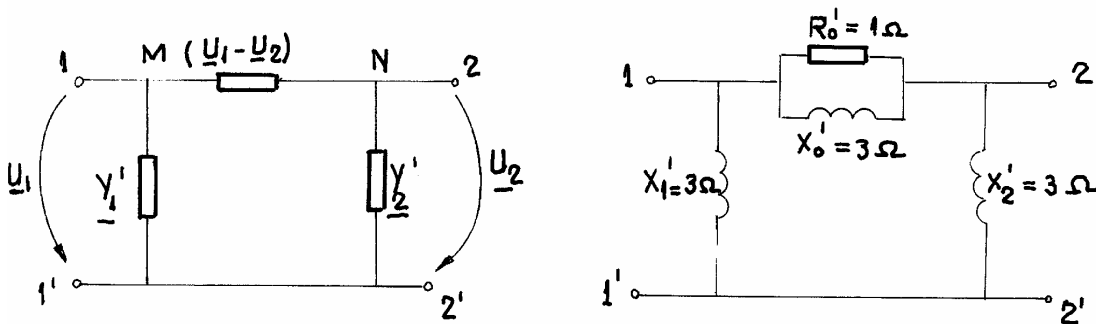


Fig. 13.4.1 a, b

de unde rezulta:

$$\underline{Y}'_1 = \underline{Y}'_{11} + \underline{Y}'_{12} = \frac{3-2j}{3} + \frac{-3+j}{3} = -\frac{j}{3} S$$

$$\underline{Y}'_2 = -(\underline{Y}'_{22} + \underline{Y}'_{21}) = \left( -\frac{-3+2j}{3} + \frac{3-j}{3} \right) = -\frac{j}{3} S$$

$$\underline{Y}'_0 = \underline{Y}'_{21} = -\underline{Y}'_{12} = -\frac{-3+j}{3} = \frac{3-j}{3} = \left( 1 - \frac{j}{3} \right) S$$

Realizarea cuadripolului in  $\pi$  echivalent este indicata in fig.13.4.1.b.

13.5 (R) Sa se determine parametrii fundamentali  $\underline{A}$ ,  $\underline{B}$ ,  $\underline{C}$ ,  $\underline{D}$  ai cuadripolilor din fig.13.5.a si 13.5.b. Pentru cuadripolul din fig.13.5.b sa se determine apoi impedantele de gol si de scurtcircuit precum si coeficientii  $\underline{Z}_{11}$ ,  $\underline{Z}_{12}$ ,  $\underline{Z}_{21}$ ,  $\underline{Z}_{22}$ .

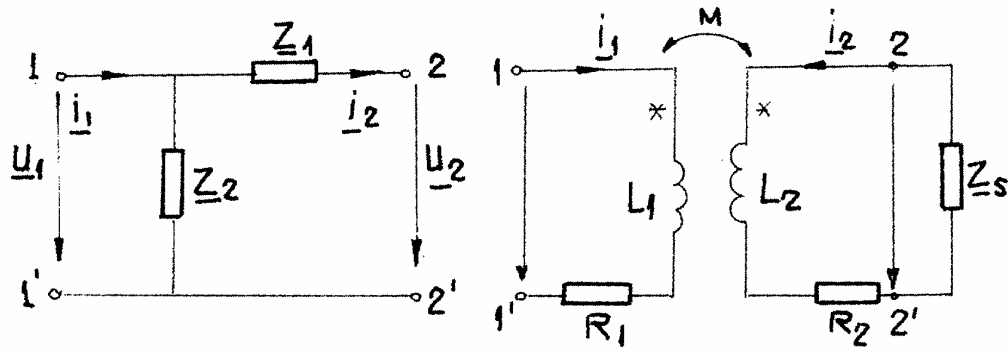


Fig. 13.5 a, b

*Rezolvare:* Pentru cuadripolul din fig.13.5.a prin incercarile de mers in gol si de scurtcircuit, cu alimentare pe la bornele primare se obtine:

$$\underline{I}_2 = 0; \underline{U}_{10} = \underline{A} \cdot \underline{U}_{20}; \underline{I}_{10} = \underline{C} \cdot \underline{U}_{20}$$

Dar  $\underline{I}_{10} = \underline{U}_{10} / \underline{Z}_2$  iar  $\underline{U}_{10} = \underline{U}_{20}$

Deci:  $\underline{A} = \underline{U}_{10} / \underline{U}_{20} = 1$ ;  $\underline{C} = \underline{I}_{10} / \underline{U}_{20} = \underline{I}_{10} / \underline{U}_{10} = 1 / \underline{Z}_2$

La scurtcircuitarea bornelor secundare:  $\underline{U}_2 = 0$  si rezulta:

$$\underline{U}_{1sc} = \underline{B} \cdot \underline{I}_{2sc}; \underline{I}_{1sc} = \underline{D} \cdot \underline{I}_{2sc}, \text{ iar din figura}$$

$$\underline{I}_{2sc} = \underline{U}_{1sc} / \underline{Z}_1; \underline{I}_{2sc} = \frac{\underline{I}_{1sc} \cdot \underline{Z}_2}{\underline{Z}_1 + \underline{Z}_2}$$

si deci:  $\underline{B} = \underline{Z}_1$ ;  $\underline{D} = \frac{\underline{Z}_1 + \underline{Z}_2}{\underline{Z}_2}$

In cazul cuadripolului din fig.13.5.b, aplicând teorema a II-a lui Kirchhoff in circuitul primar si in cel secundar al transformatorului se obtin ecuatiile:

$$\begin{aligned} \underline{U}_1(R_1 + j\omega L_1)I_1 - j\omega M \cdot I_2 \\ - \underline{U}_2 = -j\omega M \cdot I_1 + (R_2 + j\omega L_2)I_2 \end{aligned}$$

Explicitând acest sistem de ecuatii in raport cu marimile din primar, se obtin ecuatiile:

$$\underline{I}_1 = \frac{\underline{U}_2}{j\omega M} + \frac{R_2 + j\omega L_2}{j\omega M} \cdot I_2$$

$$\underline{U}_1 = \frac{\omega L_1 - jR_1}{\omega M} \underline{U}_2 + \left[ \frac{R_1 L_2}{M} + \frac{R_2 L_1}{M} + j \left( \frac{\omega L_1 L_2}{M} - \frac{R_1 R_2}{\omega M} - \omega M \right) \right] \cdot I_2 \text{ Identificând}$$

cu ecuatiile fundamentale ale cuadripolului:

$$\underline{U}_1 = \underline{A} \cdot \underline{U}_2 + \underline{B} \cdot \underline{I}_2$$

$$\underline{I}_1 = \underline{C} \cdot \underline{U}_2 + \underline{D} \cdot \underline{I}_2$$

rezulta:

$$\underline{A} = \frac{\omega L_1 - jR_1}{\omega M}; \underline{B} = \frac{R_1 L_2}{M} + \frac{R_2 L_1}{M} + j \left( \frac{\omega L_1 L_2}{M} - \frac{R_1 R_2}{\omega M} - \omega M \right)$$

$$\underline{C} = -j \cdot 1 / \omega M; \underline{D} = \frac{R_2 + j\omega L_2}{j\omega M} = L_2 / M - j \cdot R_2 / \omega M$$

Identificând primul sistem de ecuatii cu sistemul:

$$\underline{U}_1 = \underline{Z}_{11} \cdot \underline{I}_1 + \underline{Z}_{12} \cdot \underline{I}_2$$

$$\underline{U}_2 = \underline{Z}_{21} \cdot \underline{I}_1 + \underline{Z}_{22} \cdot \underline{I}_2$$

se obtine:  $\underline{Z}_{11} = R_1 + j\omega L_1$

$$\underline{Z}_{12} = -\underline{Z}_{21} = -j\omega M$$

$$\underline{Z}_{22} = -(R_2 + j\omega L_2)$$

In cazul incercarii de mers in gol ( $I_2=0$ ) ecuatiile devin:

$$\underline{U}_{10} = (R_1 + j\omega L_1) \cdot \underline{I}_{10}$$

$$\underline{U}_{20} = j\omega M \cdot \underline{I}_{10}, \text{ deci impedanta de gol este:}$$

$$\underline{Z}_{10} = \underline{U}_{10} / \underline{I}_{10} = R_1 + j\omega L_1$$

Scurtcircuitând bornele secundare ( $U_2=0$ ) se obtin ecuatiile

$$\underline{U}_{1sc} = (R_1 + j\omega L_1) \underline{I}_{1sc} - j\omega M \cdot \underline{I}_{2sc}$$

$$0 = -j\omega M \cdot \underline{I}_{1sc} + (R_2 + j\omega L_2) \cdot \underline{I}_{2sc}$$

de unde prin eliminarea lui  $\underline{I}_{2sc}$  se obtine ecuatia:

$$\underline{U}_{1sc} = \left[ R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2} \right] \underline{I}_{1sc}; \text{ impedanta de scurt-circuit, in cazul alimentarii}$$

pe la bornele primare este:

$$\underline{Z}_{1sc} = R_1 + j\omega L_1 + \frac{\omega^2 M^2}{R_2 + j\omega L_2} = j \left( \omega L_1 - \frac{\omega^2 M^2}{R_2^2 + \omega^2 L_2^2} \cdot \omega L_2 \right) + R_1 + \frac{\omega^2 M^2}{R_2^2 + \omega^2 L_2^2}$$

13.6. Sa se determine parametrii fundamentali ai cuadripolului in T (gama invers) si  $\Gamma$  (gama) pentru  $\underline{Z}_c + \underline{Z}_t$ :

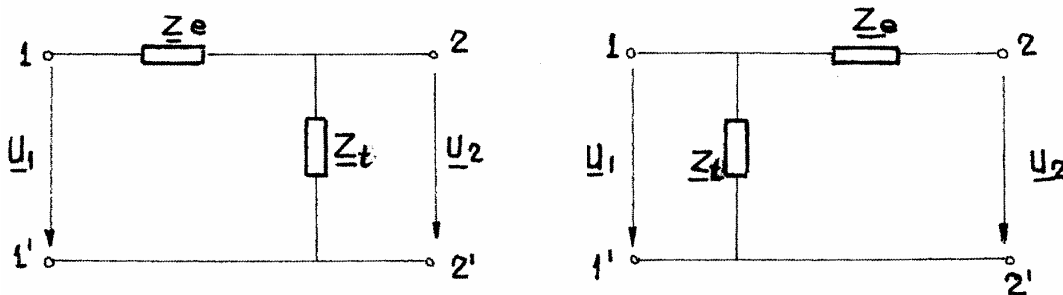


Fig. 13.6

R: a)  $\underline{A} = 1 + \underline{Z}_c / \underline{Z}_t$ ;  $\underline{B} = \underline{Z}_1$ ;  $\underline{C} = 1 / \underline{Z}_t$ ;  $\underline{D} = 1$

b)  $\underline{A} = 1$ ;  $\underline{B} = \underline{Z}_1$ ;  $\underline{C} = 1 / \underline{Z}_t$ ;  $\underline{D} = 1 + \underline{Z}_c / \underline{Z}_t$

13.7. Pentru cuadripolul din fig.13.7 se cunosc:  $\omega L_1 = \omega L_2 = 1 / \omega C = 10 \Omega$ ,  $\omega L = 5 \Omega$ . Se cere cuadripolul echivalent in T.

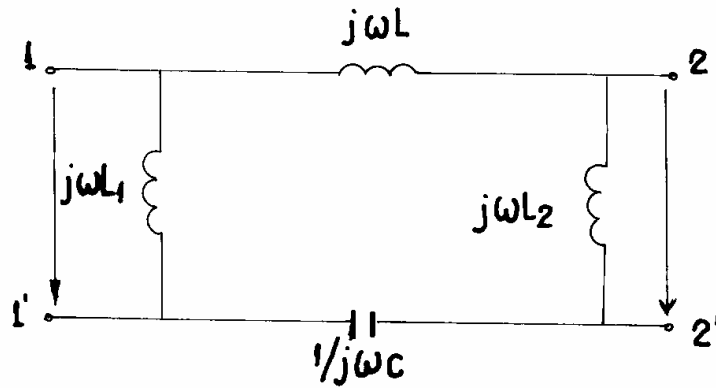


Fig. 13.7

R:  $\underline{Y} = -j \cdot 0,15 \text{ S}$ ;  $\underline{Z}_1 = -j \cdot 10/3 \Omega$ ;  $\underline{Z}_2 = -j \cdot 10/3 \Omega$ .

13.8. Se da cuadripolul din fig. 13.8 cu datele:  $\omega L_3 = 1/\omega C_3 = 10 \Omega$ ;  $R_1 = R_2 = R_3 = 1 \Omega$ ;  $\omega L_1 = \omega L_2 = 2 \Omega$ ;  $1/\omega C_1 = 1 \Omega$ . Se cere cuadripolul echivalent in  $\pi$ .

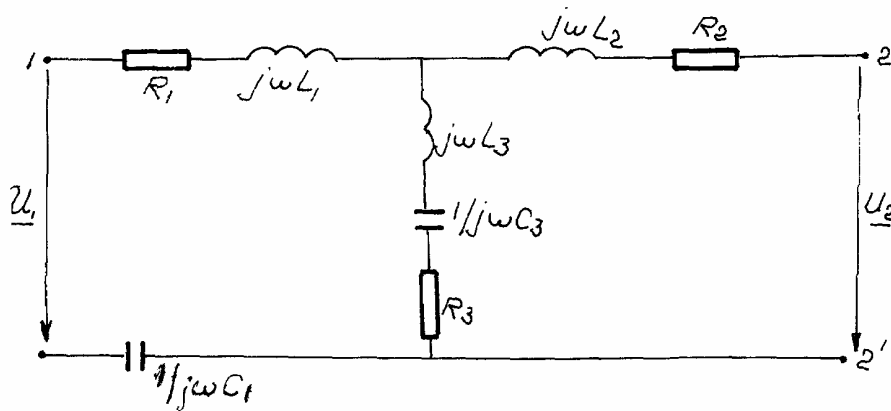


Fig.13.8

R:  $\underline{Y} = 3/2 - j$ ;  $\underline{Z}_1 = (5 - j)/13$ ;  $\underline{Z}_2 = (5 - j)/26$

13.9. Se da cuadripolul din fig.13.9 cu datele:  $\omega L_1 = 1/\omega C = 4 \Omega$ ,  $\omega L_2 = \omega L_2' = \omega L = 2 \Omega$ . Se cere cuadripolul echivalent in T.

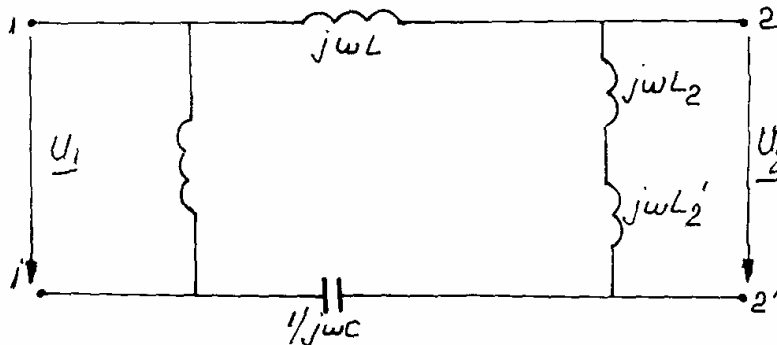


Fig. 13.9

R:  $\underline{Y} = -3/8 \cdot j$ ;  $\underline{Y}_1 = -4/3 \cdot j$ ;  $\underline{Z}_2 = -4/3 \cdot j$

13.10. Pentru cuadripolul din fig. 13.10 se cunosc:  $R_1=R_2=R_3=1 \Omega$ ;  $\omega L_1=4 \Omega$ ;  $\omega L_1'=8 \Omega$ ;  $\omega L_2=\omega L_3=\omega L_{23}=6 \Omega$ ;  $1/\omega C_2=10 \Omega$ ;  $1/\omega C_3=12 \Omega$ ;  $\omega L_{11}=3 \Omega$ . Se cer cuadripolii echivalenti in T si  $\pi$ .

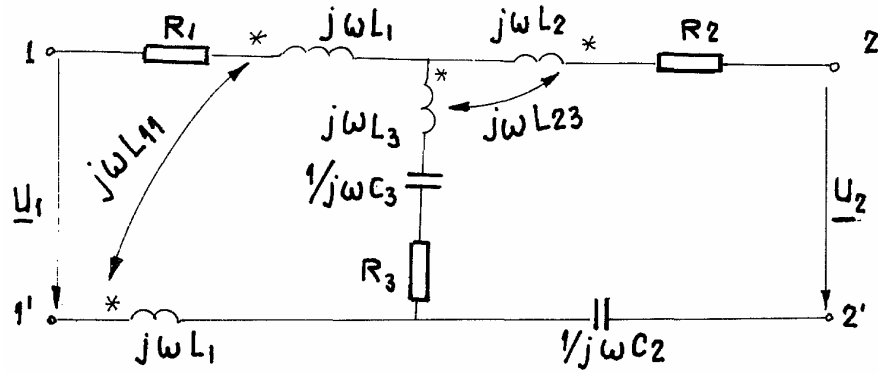


Fig. 13.10

R: a)  $\underline{Y}=1 \text{ S}$ ;  $\underline{Z}_1=1\Omega$ ;  $\underline{Z}_2=(1+2j) \Omega$

b)  $\underline{Z} = 3 + 4j\Omega$ ;  $\underline{Y}_1 = (11 + 2j)25 \cdot \text{S}$ ;  $\underline{Y}_2 = (3 - 4j)25 \cdot \text{S}$